

# Instrumentation and Controls

ETM 3301

## Lecture 22

Instructor

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## Second Order System Resonant Frequency and Resonant Peak

Gain:

$$M(\omega) = \frac{\omega_n^2}{\sqrt{(\omega_n^2 - \omega^2)^2 + (2\zeta\omega\omega_n)^2}}$$

$M(\omega)$  maximal equivalent to  $P(\omega)$  minimal.

$$P(\omega) = (\omega_n^2 - \omega^2)^2 + (2\zeta\omega\omega_n)^2$$

$$\frac{dP(\omega)}{d\omega} = 4\omega \left[ \omega^2 - (1 - 2\zeta^2)\omega_n^2 \right] = 0 \Rightarrow \omega^2 = (1 - 2\zeta^2)\omega_n^2$$

When  $1 - 2\zeta^2 > 0$  or  $\zeta < \frac{1}{\sqrt{2}} = 0.707$  A solution exists.

## Second Order System Resonant Frequency and Resonant Peak, 2

$$\omega^2 = (1 - 2\zeta^2) \omega_n^2$$

When  $\zeta < 0.707$ , the 2<sup>nd</sup> order system has resonance and the resonant frequency and resonant peak are:

Resonant frequency:  $\omega_r = \omega_n \sqrt{1 - 2\zeta^2}$

Resonant peak:  $M_p = M(\omega_r) = \frac{1}{2\zeta \sqrt{1 - \zeta^2}}$

## Resonant Frequency and Resonant Peak Example

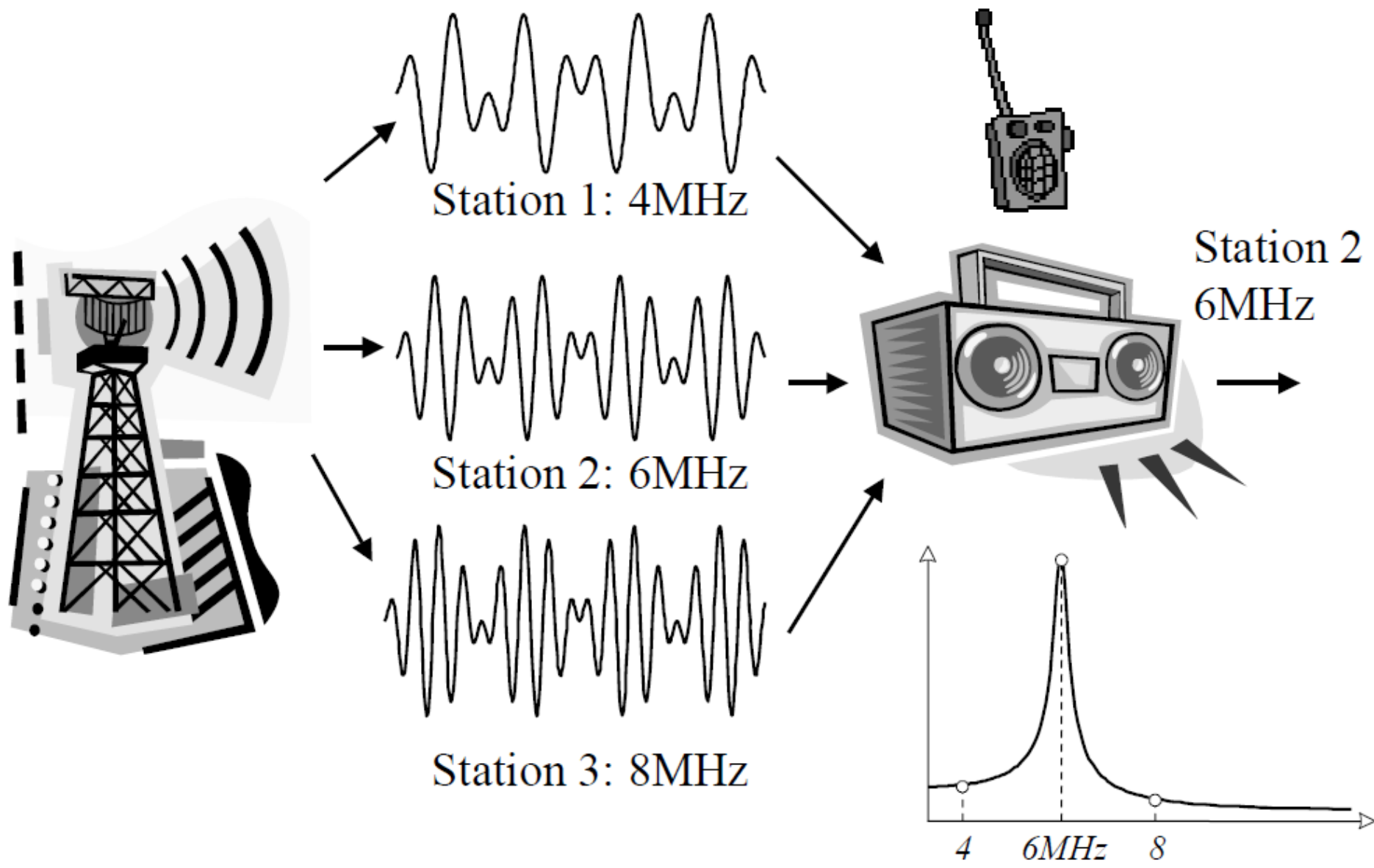
$$G(s) = \frac{25}{s^2 + 2s + 25} \quad \left. \begin{array}{l} \omega_n^2 = 25 \\ 2\zeta\omega_n = 2 \end{array} \right\} \Rightarrow \left\{ \begin{array}{l} \omega_n = 5 \\ \zeta = 0.2 \end{array} \right.$$

$\zeta = 0.2 < 0.707$ , resonance exists

Resonant frequency:  $\omega_r = \omega_n \sqrt{1 - 2\zeta^2} = 4.899 \text{ rad / s}$

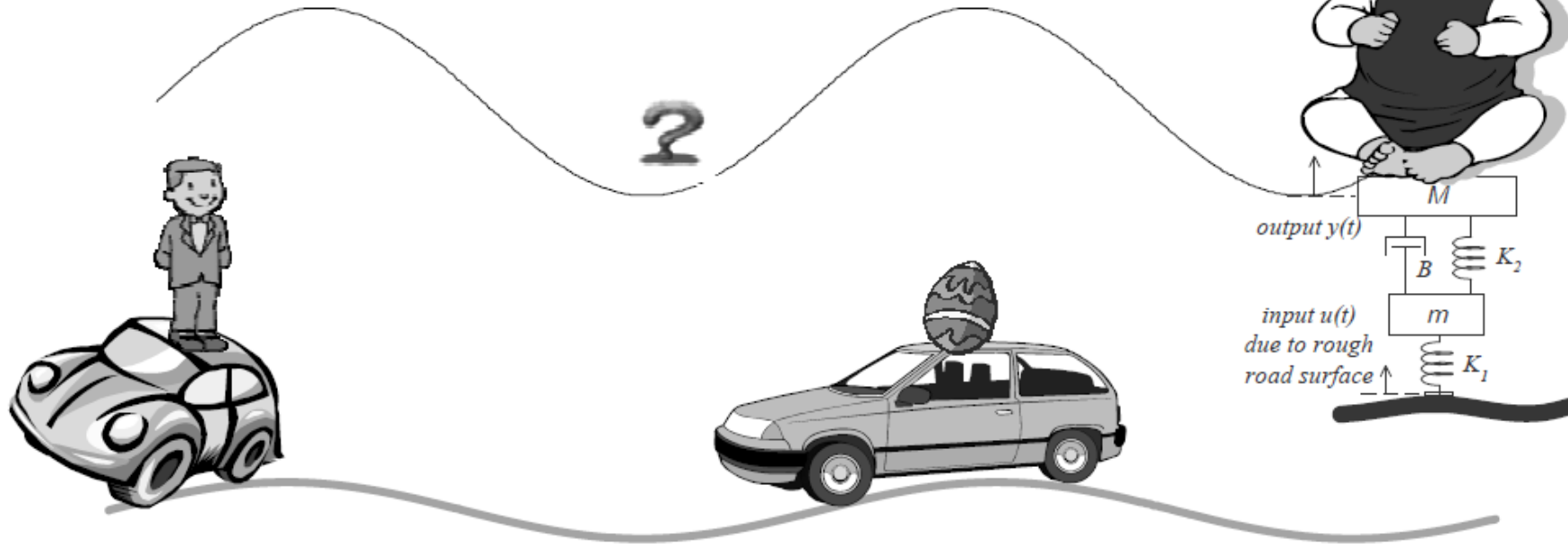
Resonant peak:  $M_p = \frac{1}{2\zeta \sqrt{1 - \zeta^2}} = 2.552$

# Resonance is "Good"



# Resonance is “Bad”

If the resonant frequency of the car’s suspension system is the same as the vibration frequency caused by bumpy road, the passengers will feel the largest vibration.



# Resonance is “Ugly (Disaster)”



External vibrations affect the bridge:

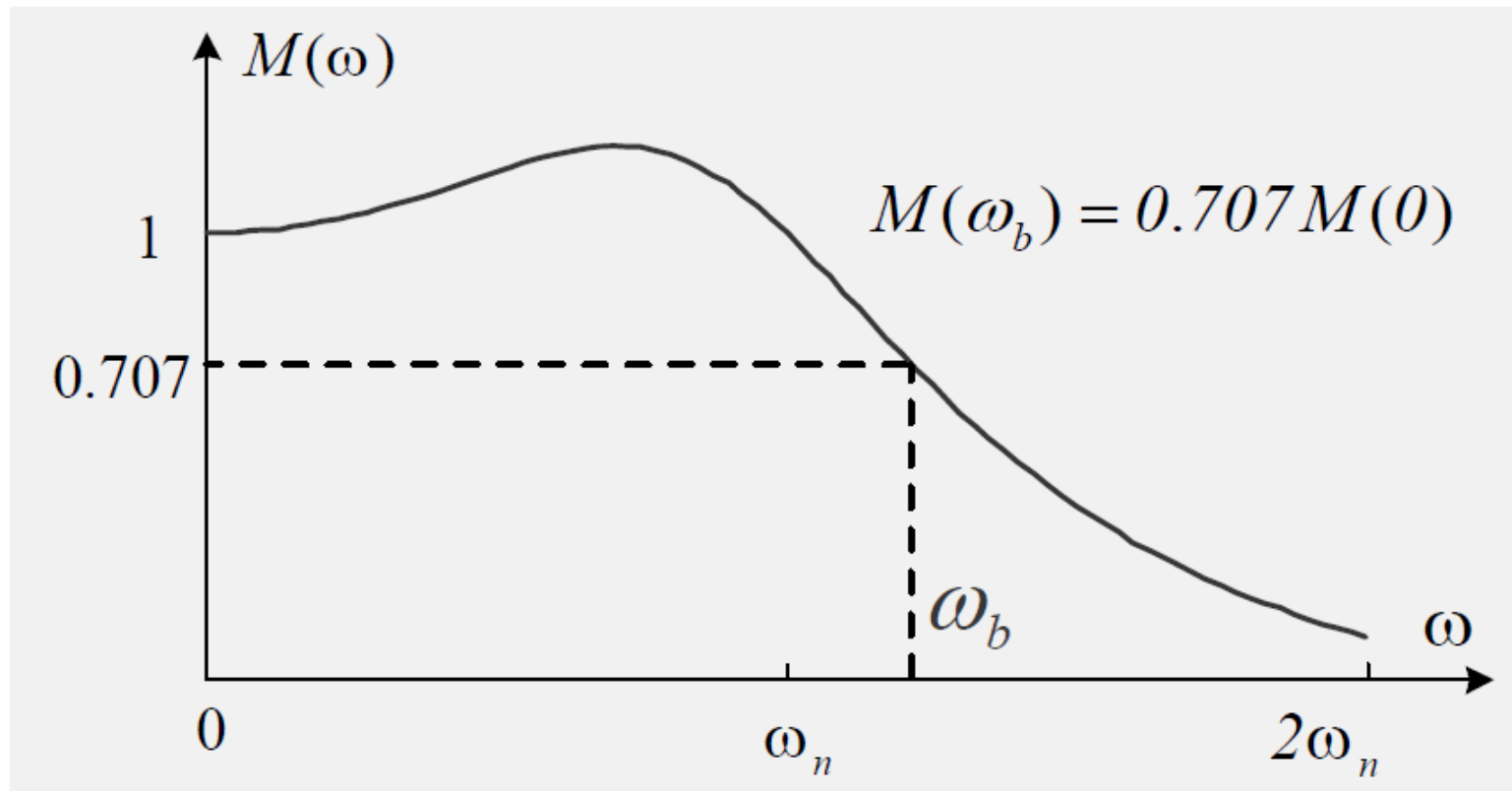
- Wind
- Movement of People and vehicles



**BBC: Swaying Millennium Bridge to close after two days:** The suspension bridge swayed alarmingly under the weight of thousands of people at its official opening on Saturday.



# Bandwidth



- **Bandwidth  $\omega_b$ :** The bandwidth  $\omega_b$  is the frequency at which the gain  $M(\omega)$  drops to 0.707 of its zero-frequency value.

2<sup>nd</sup> order system: 
$$\omega_b = \omega_n \sqrt{(1 - 2\zeta^2) + \sqrt{\zeta^4 - 4\zeta^2 + 2}}$$



## More Frequency Response Example

$$G(s) = \frac{(3s + 1)}{s(s + 1)(2s + 1)} \quad G(j\omega) = \frac{(3j\omega + 1)}{j\omega(j\omega + 1)(2j\omega + 1)}$$

Gain:

$$\begin{aligned} |G(j\omega)| &= \left| \frac{(3j\omega + 1)}{j\omega(j\omega + 1)(2j\omega + 1)} \right| = \frac{|(3j\omega + 1)|}{|j\omega(j\omega + 1)(2j\omega + 1)|} \\ &= \frac{|j3\omega + 1|}{|j\omega||j\omega + 1||j2\omega + 1|} = \frac{\sqrt{(3\omega)^2 + 1^2}}{\omega\sqrt{\omega^2 + 1^2}\sqrt{(2\omega)^2 + 1^2}} \\ &= \frac{\sqrt{9\omega^2 + 1}}{\omega\sqrt{\omega^2 + 1}\sqrt{4\omega^2 + 1}} \end{aligned}$$

## More Frequency Response Example

$$G(s) = \frac{(3s + 1)}{s(s + 1)(2s + 1)} \quad G(j\omega) = \frac{(3j\omega + 1)}{j\omega(j\omega + 1)(2j\omega + 1)}$$

Phase:

$$\angle G(j\omega) = \angle \left( \frac{(3j\omega + 1)}{j\omega(j\omega + 1)(2j\omega + 1)} \right) \quad \angle(j\omega) = 90^\circ$$

$$= \angle(3j\omega + 1) - \angle(j\omega(j\omega + 1)(2j\omega + 1))$$

$$= \angle(3j\omega + 1) - [\angle(j\omega) + \angle(j\omega + 1) + \angle(2j\omega + 1)]$$

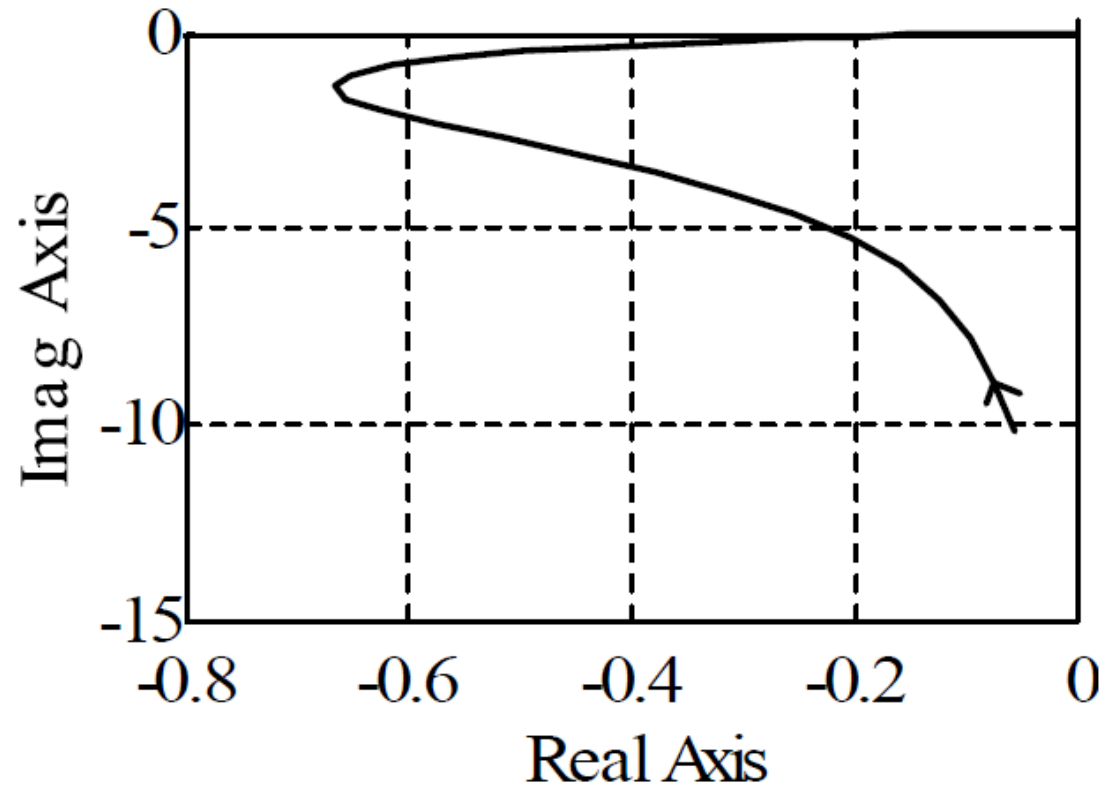
$$= \tan^{-1} \frac{3\omega}{1} - 90^\circ - \tan^{-1} \frac{\omega}{1} - \tan^{-1} \frac{2\omega}{1}$$

$$= \tan^{-1} 3\omega - 90^\circ - \tan^{-1} \omega - \tan^{-1} 2\omega$$

## More Frequency Response Example

$$|G(j\omega)| = \frac{\sqrt{9\omega^2 + 1}}{\omega\sqrt{\omega^2 + 1}\sqrt{4\omega^2 + 1}}$$

$$\angle G(j\omega) = \tan^{-1} 3\omega - 90^\circ - \tan^{-1} \omega - \tan^{-1} 2\omega$$



## Frequency response for series elements

$$G(s) = G_1(s)G_2(s)G_3(s)$$

Frequency  
response

$$G(j\omega) = G_1(j\omega)G_2(j\omega)G_3(j\omega)$$

Gain

$$|G(j\omega)| = |G_1(j\omega)||G_2(j\omega)||G_3(j\omega)|$$

Phase:

$$\angle G(j\omega) = \angle G_1(j\omega) + \angle G_2(j\omega) + \angle G_3(j\omega)$$

## Frequency response for series elements: Example

$$G(s) = \frac{1}{(s+1)(2s+1)} = \frac{1}{s+1} \times \frac{1}{2s+1} = G_1(s)G_2(s)$$

$$G_1(j\omega) = \frac{1}{j\omega + 1} \quad G_2(j\omega) = \frac{1}{j2\omega + 1}$$

$$M_1(\omega) = \frac{1}{\sqrt{\omega^2 + 1}} \quad M_2(\omega) = \frac{1}{\sqrt{4\omega^2 + 1}}$$

$$\phi_1(\omega) = -\tan^{-1} \omega \quad \phi_2(\omega) = -\tan^{-1} 2\omega$$

Gain:  $M(\omega) = M_1M_2 = \frac{1}{\sqrt{\omega^2 + 1}} \times \frac{1}{\sqrt{4\omega^2 + 1}}$

Phase:  $\phi(\omega) = \phi_1(\omega) + \phi_2(\omega) = -\tan^{-1} \omega - \tan^{-1} 2\omega$

# Gain response for series elements

$$G(s) = G_1(s)G_2(s)G_3(s)$$

$$G(j\omega) = G_1(j\omega)G_2(j\omega)G_3(j\omega)$$

Gain  $|G(j\omega)| = |G_1(j\omega)||G_2(j\omega)||G_3(j\omega)|$

Phase  $\angle G(j\omega) = \angle G_1(j\omega) + \angle G_2(j\omega) + \angle G_3(j\omega)$

Logarithmic gain

$$20\log_{10}|G(j\omega)| = 20\log_{10}|G_1(j\omega)| \\ + 20\log_{10}|G_2(j\omega)| + 20\log_{10}|G_3(j\omega)|$$

# Gain response for series elements

$$G(s) = \frac{1}{(s+1)(2s+1)} = \frac{1}{s+1} \times \frac{1}{2s+1}$$

Gain

$$20\log_{10} \frac{1}{\sqrt{\omega^2 + 1}} + 20\log_{10} \frac{1}{\sqrt{4\omega^2 + 1}}$$
$$= -10\log_{10}(\omega^2 + 1) - 10\log_{10}(4\omega^2 + 1)$$

